

# State Space-based MPC using a Kalman Filter

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B. Wayne Bequette

- Quick Review
- State Space Models
  - Unconstrained and Constrained Solutions
- Kalman Filter for Disturbance Estimation
- Summary

**PASI 2011** | Process Modeling and Optimization  
for Energy and Sustainability

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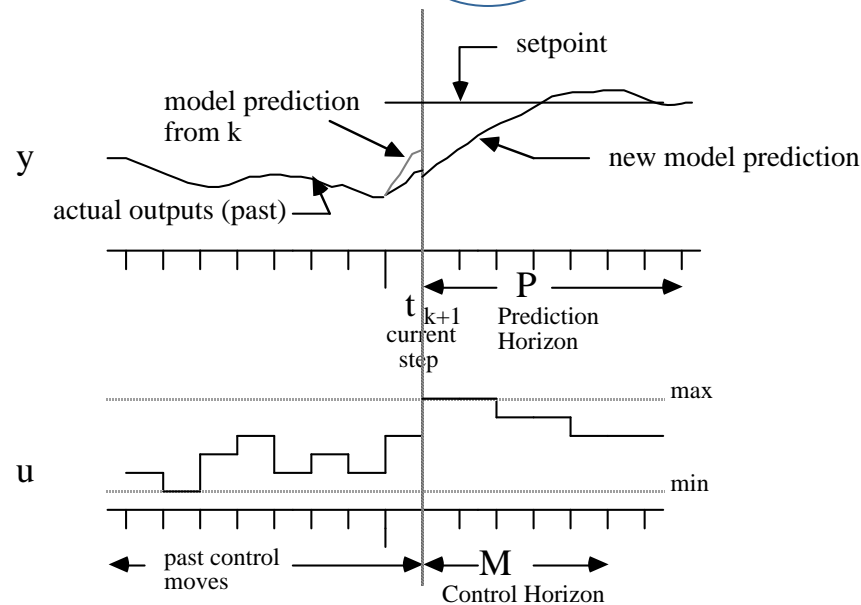
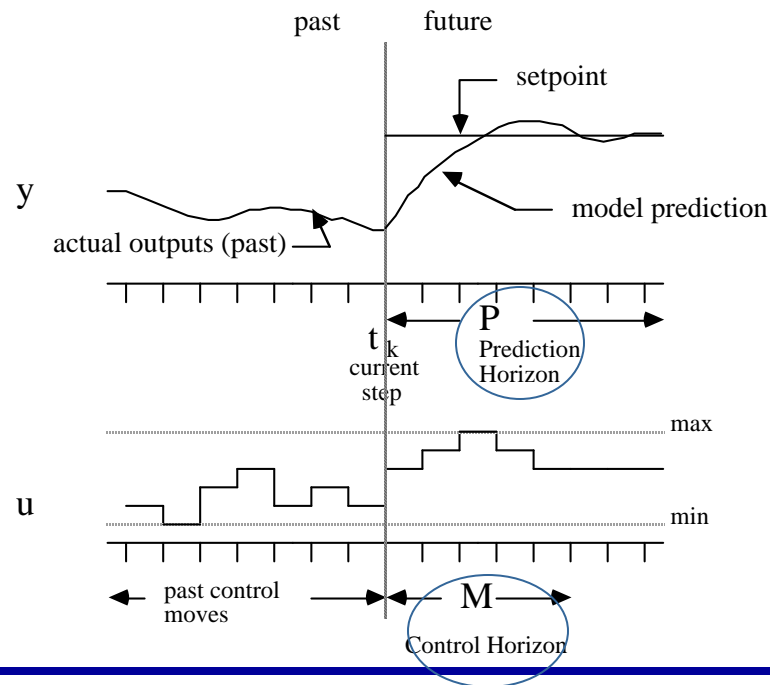


**Rensselaer**

Chemical and Biological Engineering

# Model Predictive Control (MPC)

- Constraints
- Multivariable
- Time-delays



- Quadratic objective function
- Quadratic program (QP) constraints
- State space model
- Disturbances

additive output vs. state estimation

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# Original DMC Approach to Plant-Model Mismatch

$$\hat{x}_{k|k-1} = \Phi \hat{x}_{k-1|k-1} + \Gamma u_{k-1}$$

$$\hat{y}_{k|k-1} = C \hat{x}_{k|k-1}$$

Prediction at step k, based on information at k-1

Measured output

$$\hat{p}_{k|k} = y_k - \hat{y}_{k|k-1}$$

Model output predicted from k-1

“additive output” disturbance assumption

$$\hat{y}_{k|k} = \hat{y}_{k|k-1} + \hat{p}_{k|k}$$

Forces the model “corrected output” equal to measured output

Notice that Model States are Not “Corrected”

$$\hat{x}_{k|k} = \hat{x}_{k|k-1}$$

# Model Prediction to k+1

$$\hat{x}_{k+1|k} = \Phi \hat{x}_{k|k} + \Gamma u_k$$

$$\hat{p}_{k+1|k} = \hat{p}_{k|k}$$

Assumes future corrections  
equal to current correction

$$\hat{y}_{k+1|k} = C \hat{x}_{k+1|k} + \hat{p}_{k+1|k} =$$

$$= C \Phi \hat{x}_{k|k} + C \Gamma u_k + \hat{p}_{k+1|k}$$

$$= C \Phi \hat{x}_{k|k} + C \Gamma u_{k-1} + C \Gamma \Delta u_k + \hat{p}_{k+1|k}$$

# Continue Output Predictions

$$\underbrace{\begin{bmatrix} \hat{y}_{k+1|k} \\ \hat{y}_{k+2|k} \\ \vdots \\ \hat{y}_{k+P|k} \end{bmatrix}}_{\hat{Y}} = \underbrace{\begin{bmatrix} C\Phi \\ C\Phi^2 \\ \vdots \\ C\Phi^P \end{bmatrix} \hat{x}_{k|k} + \begin{bmatrix} I \\ I \\ \vdots \\ I \end{bmatrix} \hat{p}_{k|k} + \begin{bmatrix} C\Gamma \\ C\Phi\Gamma + C\Gamma \\ \vdots \\ \sum_{i=1}^P C\Phi^{i-1}\Gamma \end{bmatrix} u_{k-1}}_{\text{"free" or "unforced response" (if no more control moves are made)}}$$

$$\begin{matrix} f \\ \\ S_f \end{matrix} \begin{bmatrix} C\Gamma & 0 & \dots & 0 \\ C\Phi\Gamma + C\Gamma & C\Gamma & 0 & 0 \\ \vdots & \vdots & & \\ \sum_{i=1}^P C\Phi^{i-1}\Gamma & \sum_{i=1}^{P-1} C\Phi^{i-1}\Gamma & \dots & \end{bmatrix} \begin{bmatrix} \Delta u_k \\ \Delta u_{k+1} \\ \vdots \\ \Delta u_{k+M-1} \end{bmatrix} \rightarrow \Delta u_f$$

"forced" response

# Output Predictions

$$\underbrace{\begin{bmatrix} \hat{y}_{k+1|k} \\ \hat{y}_{k+2|k} \\ \vdots \\ \hat{y}_{k+P|k} \end{bmatrix}}_{\hat{Y}} = \underbrace{\begin{bmatrix} C\Phi \\ C\Phi^2 \\ \vdots \\ C\Phi^P \end{bmatrix} \hat{x}_{k|k} + \begin{bmatrix} I \\ I \\ \vdots \\ I \end{bmatrix} \hat{p}_{k|k} + \begin{bmatrix} S_1 \\ S_2 \\ \vdots \\ S_P \end{bmatrix} u_{k-1}}_{\text{"free" or "unforced response" (if no more control moves are made)}}$$

$$f \begin{bmatrix} S_1 & 0 & \cdots & 0 \\ S_2 & S_1 & 0 & 0 \\ \vdots & \vdots & & \\ S_P & S_{P-1} & \cdots & S_{P-M+1} \end{bmatrix} \begin{bmatrix} \Delta u_k \\ \Delta u_{k+1} \\ \vdots \\ \Delta u_{k+M-1} \end{bmatrix}$$

"forced" response

$S_f$   $\Delta u_f$

# Same Optimization Problem as Before

$$\min_{\Delta u_f} J = \underbrace{\sum_{i=1}^P \left( r_{k+i|k} - \hat{y}_{k+i|k} \right)^T W^y \left( r_{k+i|k} - \hat{y}_{k+i|k} \right)}_{\hat{E}^T W^Y \hat{E}} + \underbrace{\sum_{i=0}^{M-1} \Delta u_{k+i}^T W^u \Delta u_{k+i}}_{\Delta u_f^T W^U \Delta u_f}$$

Where

$$W^Y = \begin{bmatrix} W^y & 0 & 0 \\ 0 & \ddots & 0 \\ 0 & 0 & W^y \end{bmatrix} \quad W^U = \begin{bmatrix} W^u & 0 & 0 \\ 0 & \ddots & 0 \\ 0 & 0 & W^u \end{bmatrix}$$

future setpoints

and

$$\hat{E} = r - \hat{Y} = \underbrace{r - f}_E - S_f \Delta u_f$$

← “unforced” error

so

$$\hat{E} = E - S_f \Delta u_f$$

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# Optimization Problem...

$$\begin{aligned}\hat{E}^T W^Y \hat{E} &= (E - S_f \Delta u_f)^T W^Y (E - S_f \Delta u_f) \\ &= E^T W^Y E - 2\Delta u_f^T S_f^T W^Y E + \Delta u_f^T S_f^T W^Y S_f \Delta u_f\end{aligned}$$

so

$$\min_{\Delta u_f} J = \hat{E}^T W^Y \hat{E} + \Delta u_f^T W^U \Delta u_f$$

can be written

$$\min_{\Delta u_f} J = E^T W^Y E + \Delta u_f^T (S_f^T W^Y S_f + W^U) \Delta u_f - 2\Delta u_f^T S_f^T W^Y E$$

and the unconstrained solution is found from

$$\partial J / \partial \Delta u_f = 0$$



# Unconstrained Solution

## Analytical Solution for Unconstrained System

$$\Delta u_f = \left( S_f^T W^Y S_f + W^U \right)^{-1} S_f^T W^Y E$$

“unforced” error



**In practice, do not actually invert a matrix. Solve as set of simultaneous equations (or use \ in MATLAB)**

$$\Delta u_f = \left( S_f^T W^Y S_f + W^U \right) \backslash S_f^T W^Y E$$

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# Vector of Control Moves

$$\Delta u_f = \begin{bmatrix} \Delta u_k \\ \Delta u_{k+1} \\ \vdots \\ \Delta u_{k+M-1} \end{bmatrix}$$

current and future moves

**Although a set of control moves is computed, only the first move  $\Delta u_k$  is implemented. The next output at  $k+1$  is obtained, then a new optimization problem is solved.**

# Problems with “Classical MPC” (e.g. DMC)

- Finite Step/Impulse Models Limited
  - Limited to open-loop stable processes (there is no corrective feedback to model states)
- Additive Output Disturbance Assumption
  - Poor performance for input step disturbances
  - No explicit measurement noise trade-off

*Ind. Eng. Chem. Res.* 2002, 41, 3745–3750

## Process Control: As Taught vs as Practiced

Francis G. Shinskey



## LIMITATIONS OF DYNAMIC MATRIX CONTROL

P. LUNDSTRÖM,<sup>1</sup>† J. H. LEE<sup>1</sup>, ‡, M. MORARI<sup>1</sup>, § and S. SKOGESTAD<sup>2</sup>

*Computers chem. Engng* Vol. 19, No. 4, pp. 409–421, 1995

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# MPC, Including Disturbance Models

$$\begin{aligned} x_{k+1} &= \Phi x_k + \Gamma u_k + \Gamma^d d_k \\ y_k &= C x_k \end{aligned}$$

(nx1)            (nxn)(nx1)    (nxn<sub>u</sub>)(n<sub>u</sub>x1)    (nxn<sub>d</sub>) (n<sub>d</sub>x1)  
(n<sub>y</sub>x1)            (n<sub>y</sub>xn)(nx1)

For perturbations  
to manipulated  
inputs

$$\Gamma^d = \Gamma$$

with known current state, easy to propagate estimates

$$x_{k+1} = \Phi x_k + \Gamma u_k + \Gamma^d d_k$$

$$y_{k+1} = C x_{k+1} = C \Phi x_k + C \Gamma u_k + C \Gamma^d d_k$$

and, using control changes

$$u_k = u_{k-1} + \Delta u_k$$

$$y_{k+1} = C \Phi x_k + C \Gamma u_{k-1} + C \Gamma^d d_k + C \Gamma \Delta u_k$$

**Next, use estimated disturbance**

# Based on an Estimated Disturbance

Predicted output                      State estimate                      Previous input                      Estimated disturbance                      Current Input change

$$\hat{y}_{k+1|k} = C\Phi\hat{x}_{k|k} + C\Gamma u_{k-1} + C\Gamma^d \hat{d}_{k|k} + C\Gamma\Delta u_k$$

Estimate at k+1                      Estimate at k                      Estimate at k

Measurements through k                      Measurements through k                      Measurements through k

Now, propagate the prediction for P steps into the future

# Output Predictions

$$\underbrace{\begin{bmatrix} \hat{y}_{k+1|k} \\ \hat{y}_{k+2|k} \\ \vdots \\ \hat{y}_{k+P|k} \end{bmatrix}}_{\hat{Y}} = \underbrace{\begin{bmatrix} C\Phi \\ C\Phi^2 \\ \vdots \\ C\Phi^P \end{bmatrix} \hat{x}_{k|k} + \begin{bmatrix} C\Gamma \\ C\Phi\Gamma + C\Gamma \\ \vdots \\ \sum_{i=1}^P C\Phi^{i-1}\Gamma \end{bmatrix} u_{k-1} + \begin{bmatrix} C\Gamma^d \\ C\Phi\Gamma^d + C\Gamma^d \\ \vdots \\ \sum_{i=1}^P C\Phi^{i-1}\Gamma^d \end{bmatrix} \hat{d}_{k|k}}_{\text{"free" or "unforced response" (if no more control moves are made)}}$$

Estimated disturbance  
(constant in future)

$$f \begin{bmatrix} C\Gamma & 0 & \dots & 0 \\ C\Phi\Gamma + C\Gamma & C\Gamma & 0 & 0 \\ \vdots & \vdots & & \\ \sum_{i=1}^P C\Phi^{i-1}\Gamma & \sum_{i=1}^{P-1} C\Phi^{i-1}\Gamma & \dots & \end{bmatrix} \begin{bmatrix} \Delta u_k \\ \Delta u_{k+1} \\ \vdots \\ \Delta u_{k+M-1} \end{bmatrix}$$

$S_f$       "forced" response       $\Delta u_f$

# Output Predictions (using S notation)

$$\underbrace{\begin{bmatrix} \hat{y}_{k+1|k} \\ \hat{y}_{k+2|k} \\ \vdots \\ \hat{y}_{k+P|k} \end{bmatrix}}_{\hat{Y}} = \underbrace{\begin{bmatrix} C\Phi \\ C\Phi^2 \\ \vdots \\ C\Phi^P \end{bmatrix} \hat{x}_{k|k} + \begin{bmatrix} S_1 \\ S_2 \\ \vdots \\ S_P \end{bmatrix} u_{k-1} + \begin{bmatrix} S_1^d \\ S_2^d \\ \vdots \\ S_P^d \end{bmatrix} \hat{d}_{k|k}}_{\text{"free" or "unforced response" (if no more control moves are made)}}$$

$\hat{Y}$  is labeled  $f$ .  $\hat{d}_{k|k}$  is labeled **Estimated disturbance (constant in future)**.

$$\underbrace{\begin{bmatrix} S_1 & 0 & \cdots & 0 \\ S_2 & S_2 & 0 & 0 \\ \vdots & \vdots & & \\ S_P & S_{P-1} & \cdots & S_{P-M+1} \end{bmatrix} \begin{bmatrix} \Delta u_k \\ \Delta u_{k+1} \\ \vdots \\ \Delta u_{k+M-1} \end{bmatrix}}_{\text{"forced" response}}$$

$f$  is labeled  $S_f$ .  $\Delta u_f$  is labeled  $\Delta u_f$ .

# Same Optimization Problem as Before

$$\min_{\Delta u_f} J = \underbrace{\sum_{i=1}^P \left( r_{k+i|k} - \hat{y}_{k+i|k} \right)^T W^y \left( r_{k+i|k} - \hat{y}_{k+i|k} \right)}_{\hat{E}^T W^Y \hat{E}} + \underbrace{\sum_{i=0}^{M-1} \Delta u_{k+i}^T W^u \Delta u_{k+i}}_{\Delta u_f^T W^U \Delta u_f}$$

Where

$$W^Y = \begin{bmatrix} W^y & 0 & 0 \\ 0 & \ddots & 0 \\ 0 & 0 & W^y \end{bmatrix} \quad W^U = \begin{bmatrix} W^u & 0 & 0 \\ 0 & \ddots & 0 \\ 0 & 0 & W^u \end{bmatrix}$$

future setpoints

and

$$\hat{E} = r - \hat{Y} = \underbrace{r - f}_E - S_f \Delta u_f$$

← “unforced” error

so

$$\hat{E} = E - S_f \Delta u_f$$



## Optimization Problem...

$$\begin{aligned}\hat{E}^T W^Y \hat{E} &= (E - S_f \Delta u_f)^T W^Y (E - S_f \Delta u_f) \\ &= E^T W^Y E - 2\Delta u_f^T S_f^T W^Y E + \Delta u_f^T S_f^T W^Y S_f \Delta u_f\end{aligned}$$

so

$$\min_{\Delta u_f} J = \hat{E}^T W^Y \hat{E} + \Delta u_f^T W^U \Delta u_f$$

can be written

$$\min_{\Delta u_f} J = E^T W^Y E + \Delta u_f^T (S_f^T W^Y S_f + W^U) \Delta u_f - 2\Delta u_f^T S_f^T W^Y E$$

and the unconstrained solution is found from

$$\partial J / \partial \Delta u_f = 0$$

# Unconstrained Solution

## Analytical Solution for Unconstrained System

$$\Delta u_f = \left( S_f^T W^Y S_f + W^U \right)^{-1} S_f^T W^Y E$$

“unforced” error

**In practice, do not actually invert a matrix. Solve as set of simultaneous equations (or use \ in MATLAB)**

$$\Delta u_f = \left( S_f^T W^Y S_f + W^U \right) \backslash S_f^T W^Y E$$

How do we estimate the disturbance?

# Concise Review of Optimal Estimation (Kalman Filtering)



**Rudy Kalman**

- Kalman & Bucy, 1960

Kalman, R. E. (1960b). A new approach to linear filtering and prediction problems. *Transactions of ASME, Journal of Basic Engineering*, 87, 35–45.

- Developed for well-modeled systems with state input and measurement noise
- Success in Aerospace and other applications
- Problems with initial chemical process applications
  - Limited understanding of limitations
- Bias due to model uncertainty
- Need an “appended state” formulation

# Stochastic Models

$$x_{k+1} = \Phi x_k + \Gamma u_k + \Gamma^w w_k \quad \leftarrow \text{Input noise} \quad \text{cov}(w) = Q$$

$$y_k = C x_k + v_k \quad \leftarrow \text{Measurement noise} \quad \text{cov}(v) = R$$

$$\text{cov}(x_0) = P_0 \quad \leftarrow \text{Initial State Uncertainty}$$

# Optimal State Estimate

$$J(\hat{y}) = \sum_{i=1}^k (y_i - C\hat{x})^T R^{-1} (y_i - C\hat{x})$$

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# State Estimation Procedure

$$\hat{\mathbf{x}}_{k|k-1} = \Phi \hat{\mathbf{x}}_{k-1|k-1} + \Gamma \mathbf{u}_{k-1} \quad \text{Prediction}$$

← Estimate at k, based on measurements through k-1

$$\hat{\mathbf{x}}_{k|k} = \hat{\mathbf{x}}_{k|k-1} + \mathbf{L}_k \left( \mathbf{y}_k - \mathbf{C} \hat{\mathbf{x}}_{k|k-1} \right) \quad \text{Correction}$$

← Estimate at k, based on measurements through k

$$\mathbf{P}_k = \Phi \mathbf{P}_{k-1} \Phi^T + \Gamma^w \mathbf{Q} \Gamma^{wT} - \Phi \mathbf{P}_{k-1} \mathbf{C}^T \left( \mathbf{C} \mathbf{P}_{k-1} \mathbf{C}^T + \mathbf{R} \right)^{-1} \mathbf{C} \mathbf{P}_{k-1} \Phi^T$$

← State error covariance estimate

Kalman gain

$$\mathbf{L}_k = \mathbf{P}_k \mathbf{C}^T \left( \mathbf{C} \mathbf{P}_k \mathbf{C}^T + \mathbf{R} \right)^{-1}$$

Discuss steady-state Kalman gain

# State Covariance Comparison

“Open-loop”

$$P_k = \Phi P_{k-1} \Phi^T + \Gamma^w Q \Gamma^{wT}$$

With measurement (Kalman Filter formulation)

$$P_k = \Phi P_{k-1} \Phi^T + \Gamma^w Q \Gamma^{wT} - \underbrace{\Phi P_{k-1} C^T (C P_{k-1} C^T + R)^{-1} C P_{k-1}}_{\text{Measurement update reduces state estimate covariance}} \Phi^T$$

Measurement update reduces state estimate covariance



# Kalman Filter w/Augmented States

$$\underbrace{\begin{bmatrix} x_{k+1} \\ d_{k+1} \end{bmatrix}}_{x_{k+1}^a} = \underbrace{\begin{bmatrix} \Phi & \Gamma^d \\ 0 & I \end{bmatrix}}_{\Phi^a} \underbrace{\begin{bmatrix} x_k \\ d_k \end{bmatrix}}_{x_k^a} + \underbrace{\begin{bmatrix} \Gamma \\ 0 \end{bmatrix}}_{\Gamma^a} u_k + \underbrace{\begin{bmatrix} 0 \\ I \end{bmatrix}}_{\Gamma^{a,w}} w_k$$

process noise  
 $\text{cov}(w) = Q$

$$y_k = \underbrace{\begin{bmatrix} C & 0 \end{bmatrix}}_{C^a} \underbrace{\begin{bmatrix} x_k \\ d_k \end{bmatrix}}_{x_k^a} + v_k$$

Augmented state (includes disturbance)  
 Measurement noise  $\text{cov}(v) = R$

## Predictor-corrector equations:

$$\hat{x}_{k|k-1}^a = \Phi^a \hat{x}_{k-1|k-1}^a + \Gamma^a u_{k-1}$$

Manipulated input

$$\hat{x}_{k|k}^a = \hat{x}_{k|k-1}^a + L_k (y_k - C^a \hat{x}_{k|k-1}^a)$$

Aug. state estimate

Kalman gain

Measured output

# Disturbance Portion of Augmented State

$$\hat{\mathbf{x}}_{k|k}^a = \begin{bmatrix} \hat{\mathbf{x}}_{k|k} \\ \hat{\mathbf{d}}_{k|k} \end{bmatrix}$$

## Constant Disturbance Into the Future

$$\hat{\mathbf{d}}_{k+j|k} = \hat{\mathbf{d}}_{k|k}$$

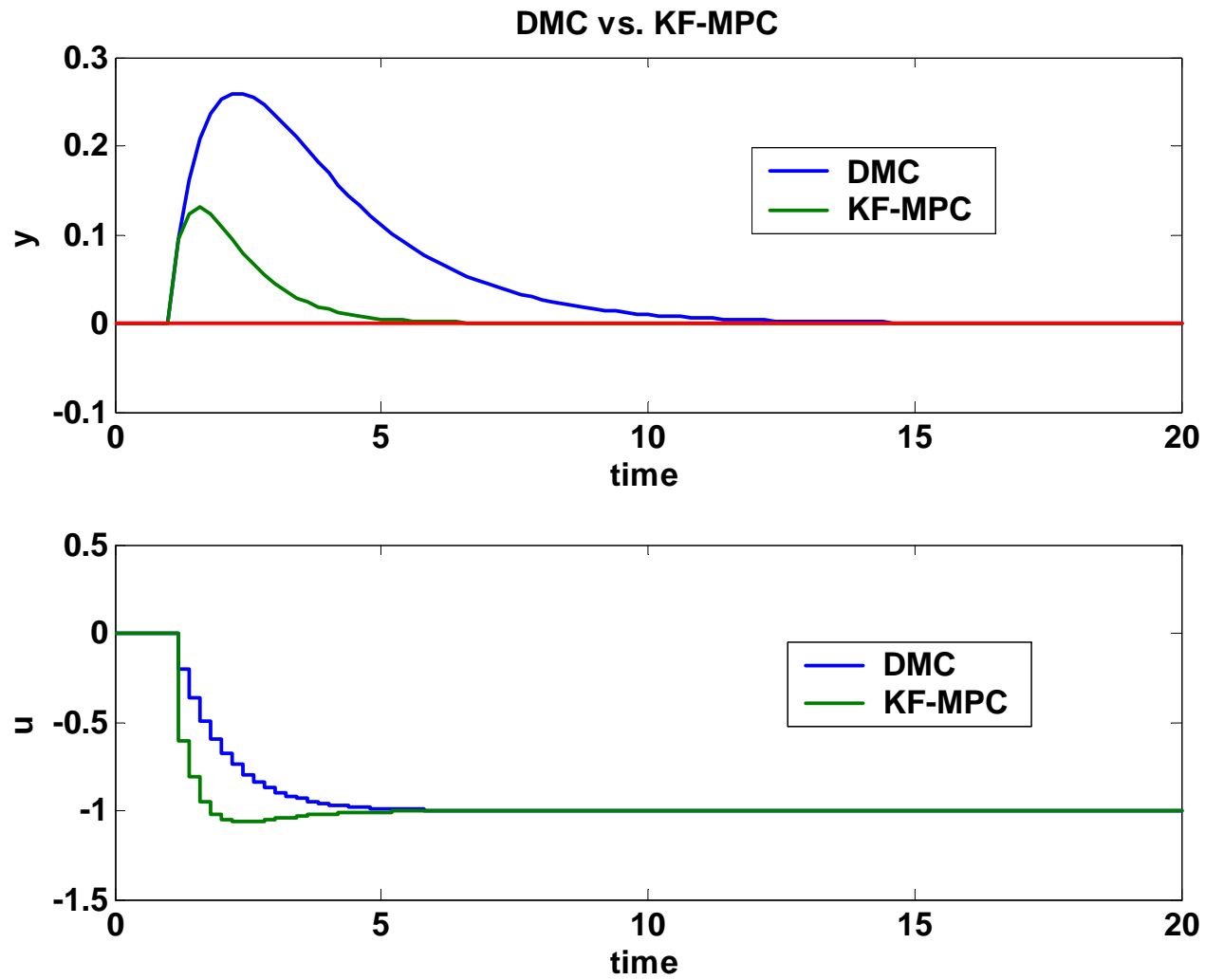
Can use physical knowledge to make better assumptions (periodic effects, etc.)

## Example

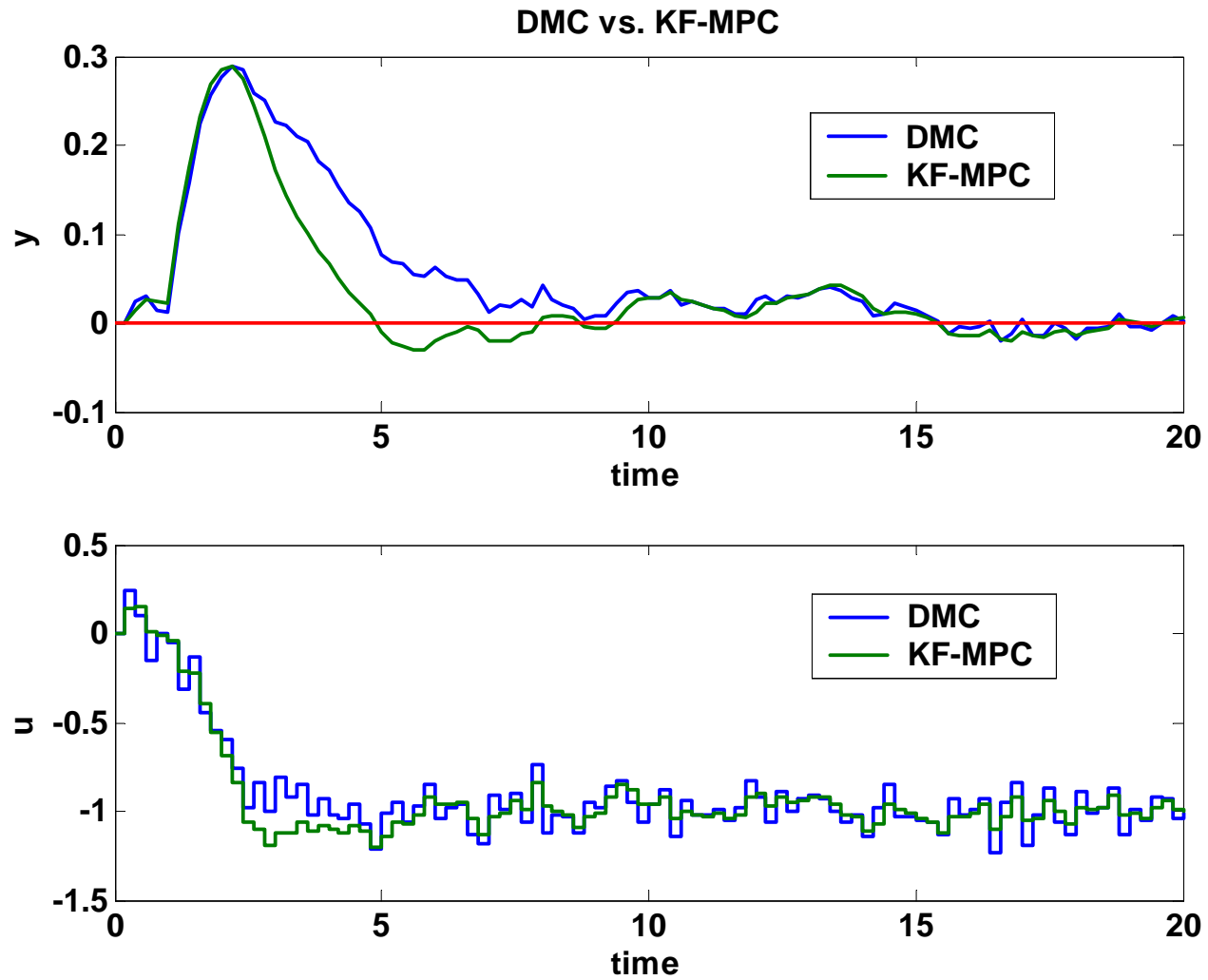
$$g_p(s) = \frac{1}{10s + 1}$$

**Comparison of DMC and KF-based MPC for a step input disturbance**

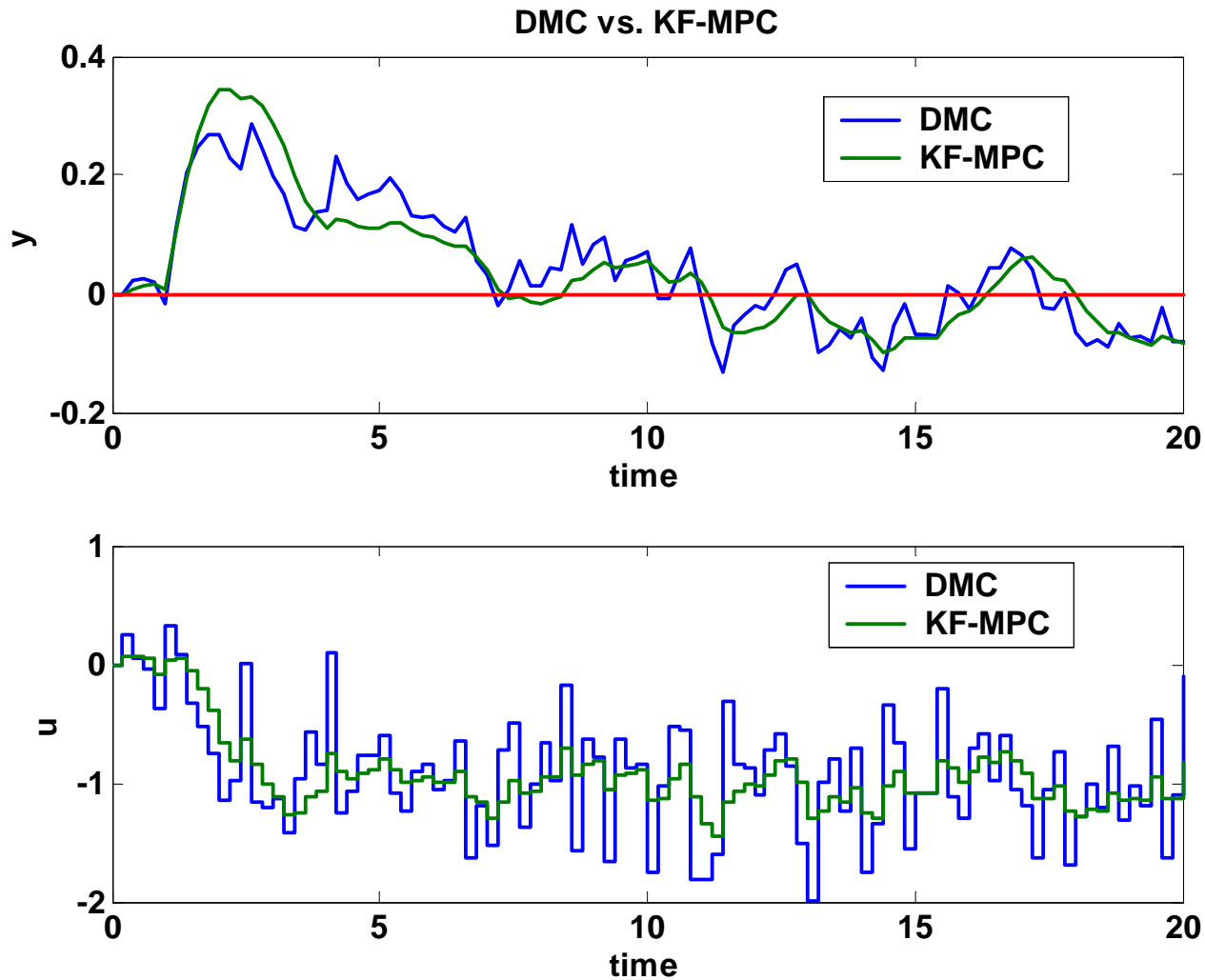
# Step Input Disturbance



# Measurement Noise



# Substantial Measurement Noise



# Constrained MPC

## Input constraints

$$\underset{(n_u \times 1)}{u_{\min}} \leq \underset{(n_u \times 1)}{u_{k+i}} \leq \underset{(n_u \times 1)}{u_{\max}}$$

where

$$u_k = u_{k-1} + \Delta u_k$$
$$u_{k+1} = u_{k-1} + \Delta u_k + \Delta u_{k+1} \quad \text{etc.}$$

## Input “velocity” constraints

$$\Delta u_{\min} \leq \Delta u_{k+i} \leq \Delta u_{\max}$$

## Output constraints

$$y_{\min} \leq \hat{y}_{k+i|k} \leq y_{\max}$$

# MPC: Constraint Formulation

Input

$$\begin{bmatrix} u_{\min} \\ u_{\min} \\ \vdots \\ u_{\min} \end{bmatrix} \leq \begin{bmatrix} u_{k-1} \\ u_{k-1} \\ \vdots \\ u_{k-1} \end{bmatrix} + \begin{bmatrix} I & 0 & 0 & \cdots & 0 \\ I & I & 0 & \cdots & 0 \\ \vdots & \vdots & & & \\ I & I & I & \cdots & I \end{bmatrix} \begin{bmatrix} \Delta u_k \\ \Delta u_{k+1} \\ \vdots \\ \Delta u_{k+M-1} \end{bmatrix} \leq \begin{bmatrix} u_{\max} \\ u_{\max} \\ \vdots \\ u_{\max} \end{bmatrix}$$

Velocity

$$\begin{bmatrix} \Delta u_{\min} \\ \Delta u_{\min} \\ \vdots \\ \Delta u_{\min} \end{bmatrix} \leq \begin{bmatrix} \Delta u_k \\ \Delta u_{k+1} \\ \vdots \\ \Delta u_{k+M-1} \end{bmatrix} \leq \begin{bmatrix} \Delta u_{\max} \\ \Delta u_{\max} \\ \vdots \\ \Delta u_{\max} \end{bmatrix}$$



# Output constraints

$$y_{\min} - f \leq S_f \Delta u_f \leq y_{\max} - f$$

**“Free” or “unforced” response**

$$f = \underbrace{\begin{bmatrix} C\Phi \\ C\Phi^2 \\ \vdots \\ C\Phi^P \end{bmatrix} \hat{x}_{k|k} + \begin{bmatrix} C\Gamma \\ C\Phi\Gamma + C\Gamma \\ \vdots \\ \sum_{i=1}^P C\Phi^{i-1}\Gamma \end{bmatrix} u_{k-1} + \begin{bmatrix} C\Gamma^d \\ C\Phi\Gamma^d + C\Gamma^d \\ \vdots \\ \sum_{i=1}^P C\Phi^{i-1}\Gamma^d \end{bmatrix} \hat{d}_{k|k}}_{\text{"free" or "unforced response" (if no more control moves are made)}}$$

# QP Codes: “one-sided” form

$$\underbrace{\begin{bmatrix} I & 0 & 0 & \cdots & 0 \\ I & I & 0 & \cdots & 0 \\ \vdots & \vdots & & & \\ I & I & I & \cdots & I \end{bmatrix}}_{\mathbf{A}_1} \underbrace{\begin{bmatrix} \Delta u_k \\ \Delta u_{k+1} \\ \vdots \\ \Delta u_{k+M-1} \end{bmatrix}}_{\Delta \mathbf{u}_f} \geq \underbrace{\begin{bmatrix} u_{\min} - u_{k-1} \\ u_{\min} - u_{k-1} \\ \vdots \\ u_{\min} - u_{k-1} \end{bmatrix}}_{\mathbf{b}_1}$$
  

$$- \underbrace{\begin{bmatrix} I & 0 & 0 & \cdots & 0 \\ I & I & 0 & \cdots & 0 \\ \vdots & \vdots & & & \\ I & I & I & \cdots & I \end{bmatrix}}_{\mathbf{A}_2} \underbrace{\begin{bmatrix} \Delta u_k \\ \Delta u_{k+1} \\ \vdots \\ \Delta u_{k+M-1} \end{bmatrix}}_{\Delta \mathbf{u}_f} \geq \underbrace{\begin{bmatrix} u_{k-1} - u_{\max} \\ u_{k-1} - u_{\max} \\ \vdots \\ u_{k-1} - u_{\max} \end{bmatrix}}_{\mathbf{b}_2}$$

“one-sided” form for input constraints

$$\begin{bmatrix} A_1 \\ A_2 \end{bmatrix} \Delta u_f \geq \begin{bmatrix} b_1 \\ b_2 \end{bmatrix}$$

“one-sided” form for output constraints

$$S_f \Delta u_f \geq y_{\min} - f$$

$$-S_f \Delta u_f \geq -y_{\max} + f$$

$$\begin{bmatrix} A_3 \\ A_4 \end{bmatrix} \Delta u_f \geq \begin{bmatrix} b_3 \\ b_4 \end{bmatrix}$$

Combining all constraints

$$\begin{bmatrix} A_1 \\ A_2 \\ A_3 \\ A_4 \end{bmatrix} \Delta u_f \geq \begin{bmatrix} b_1 \\ b_2 \\ b_3 \\ b_4 \end{bmatrix}$$

$$A \Delta u_f \geq b$$

# Quadratic Program (QP)

**Objective Function**  $\min_{\Delta u_f} \Phi = \frac{1}{2} \cdot \Delta u_f^T H \Delta u_f + c^T \Delta u_f$

**Decision variables**

**s.t.**

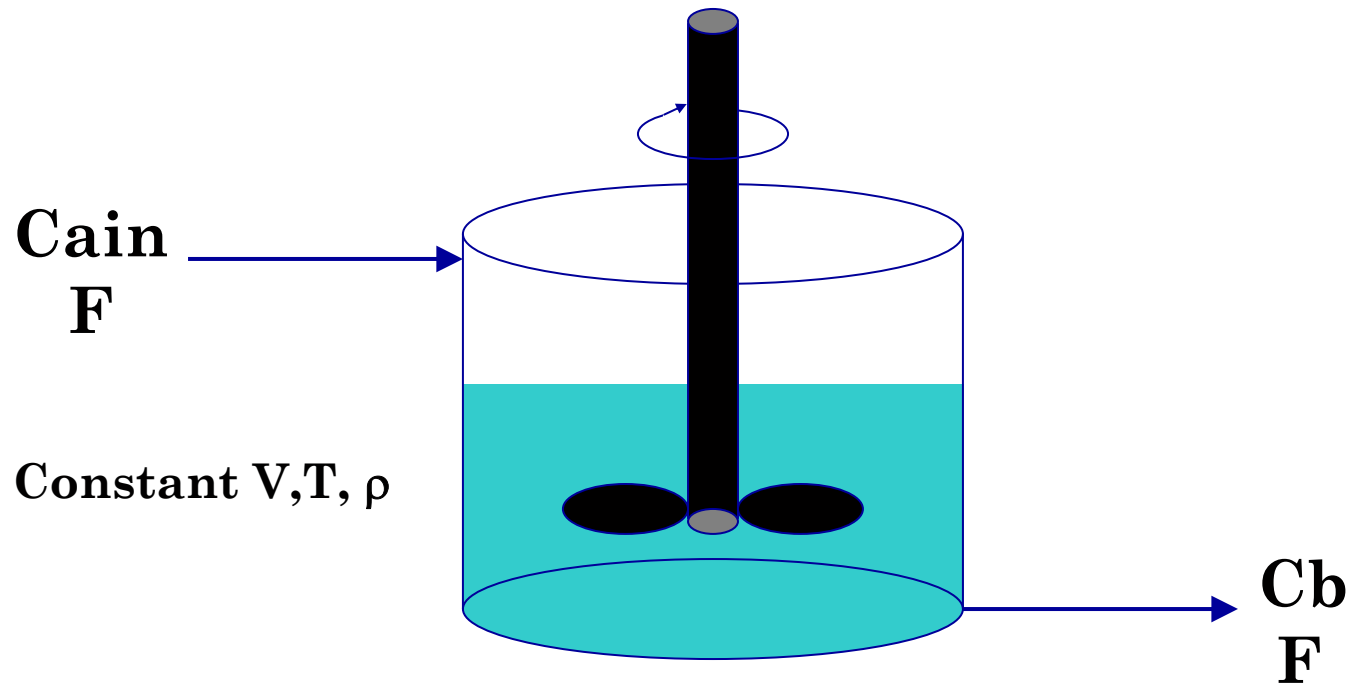
$$A \Delta u_f \geq b$$

$$\Delta u_{\min} \leq \Delta u_f \leq \Delta u_{\max}$$

**Where**  $H = S_f^T W^Y S_f + W^U$

$$c = -S_f^T W^Y E$$

# Inverse Response Process: Van de Vusse Reaction

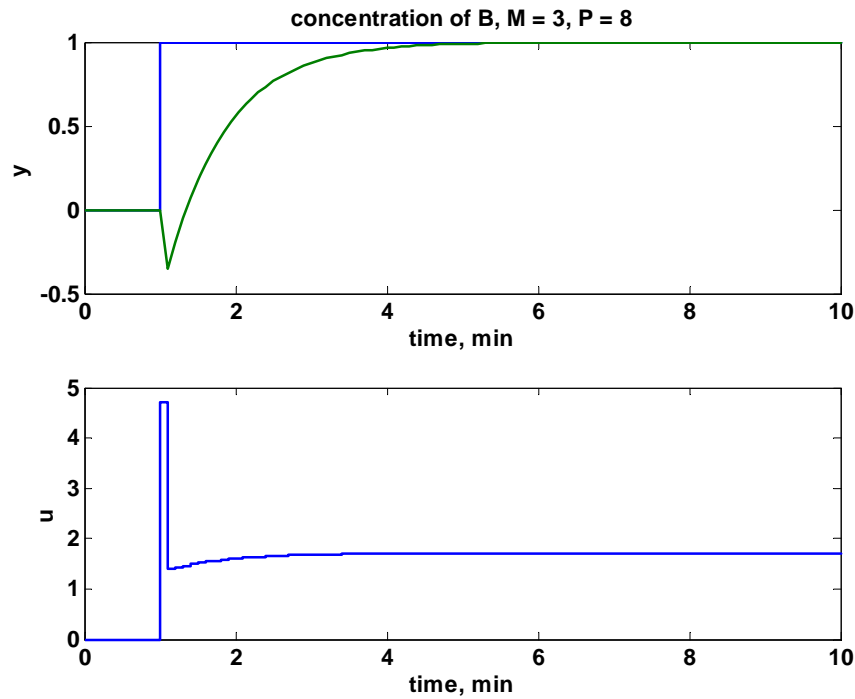


$$\frac{dC_a}{dt} = -k_1 C_a - k_3 C_a^2 + (C_{ain} - C_a)u$$

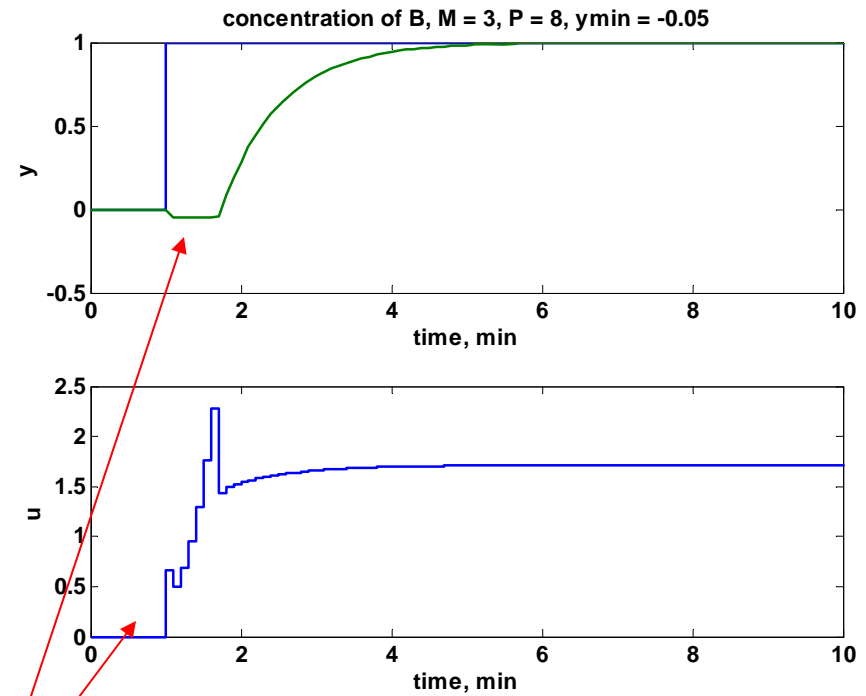
$$\frac{dC_b}{dt} = k_1 C_a - k_2 C_b - C_b u \quad \text{where } u = F/V$$

# Output Constraints: Inverse Response Example

## Unconstrained



## $y \geq -0.05$

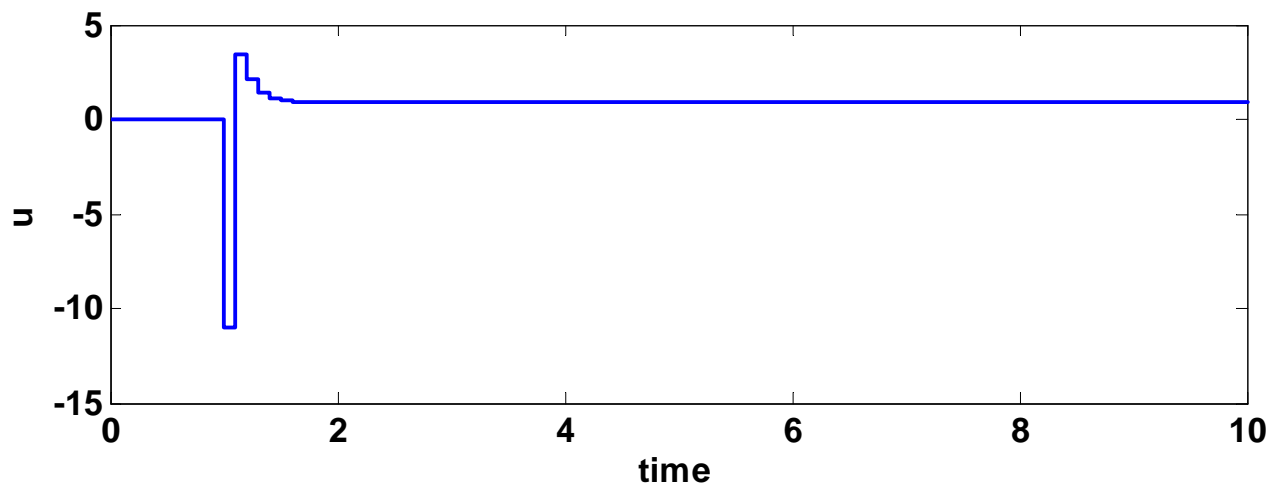
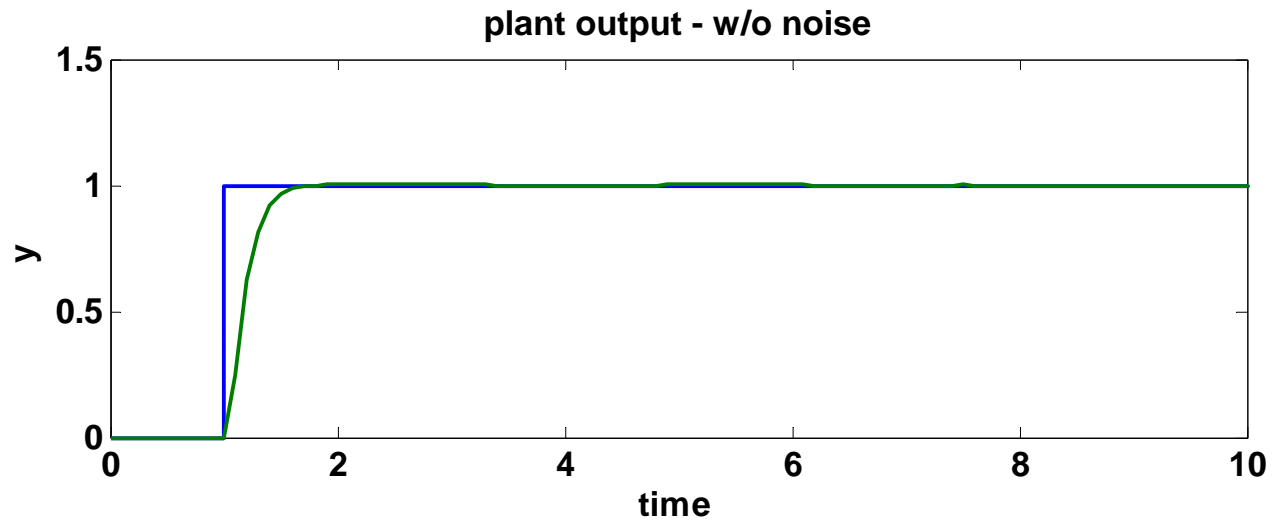


**Less control action in order to meet output constraint**

# Output Constraints

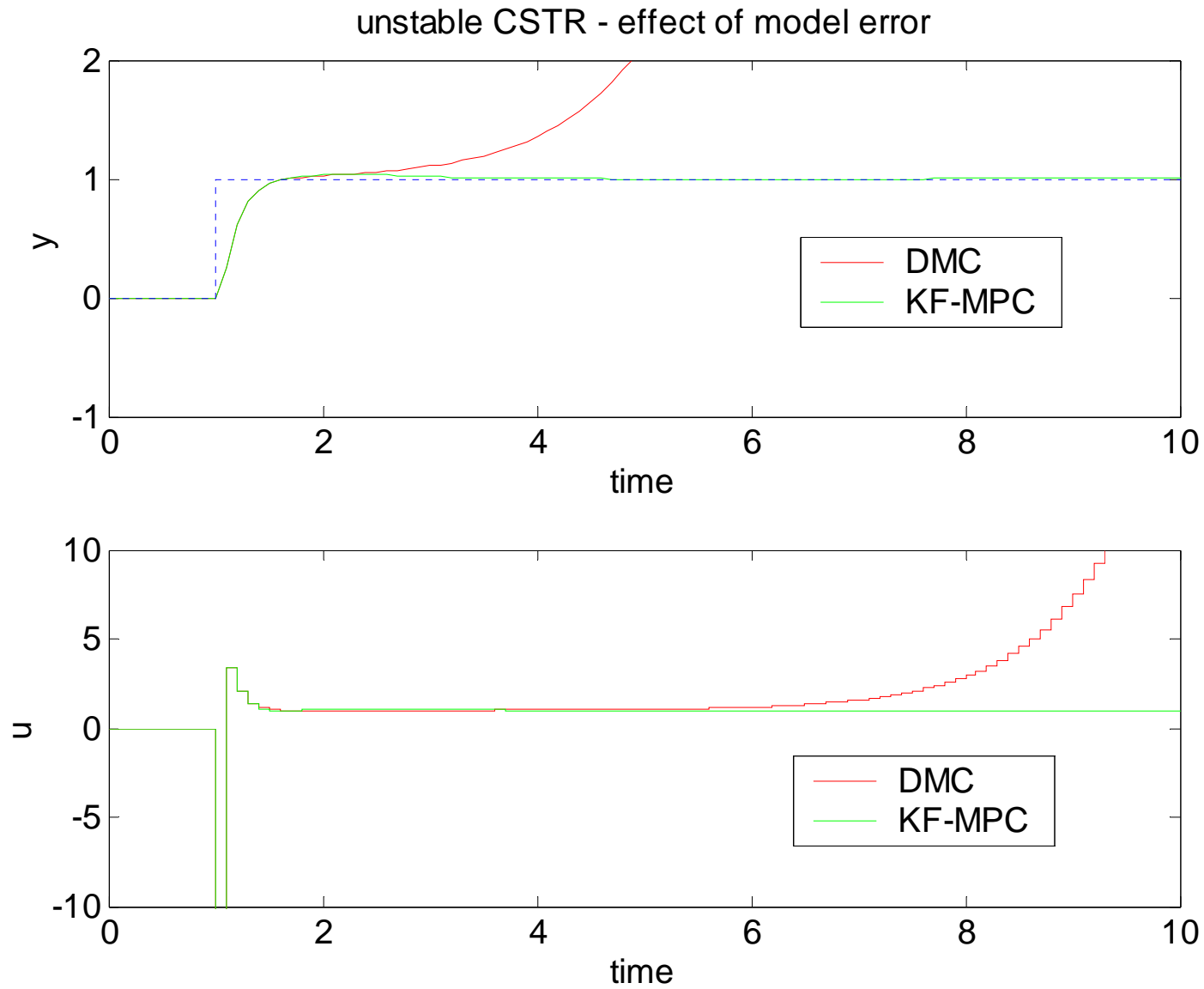
- May not exist a feasible solution
- Relax constraints until feasible
- “Soft constraints” – Penalize in Objective Function

# Unstable CSTR Example





# Effect of Model Error



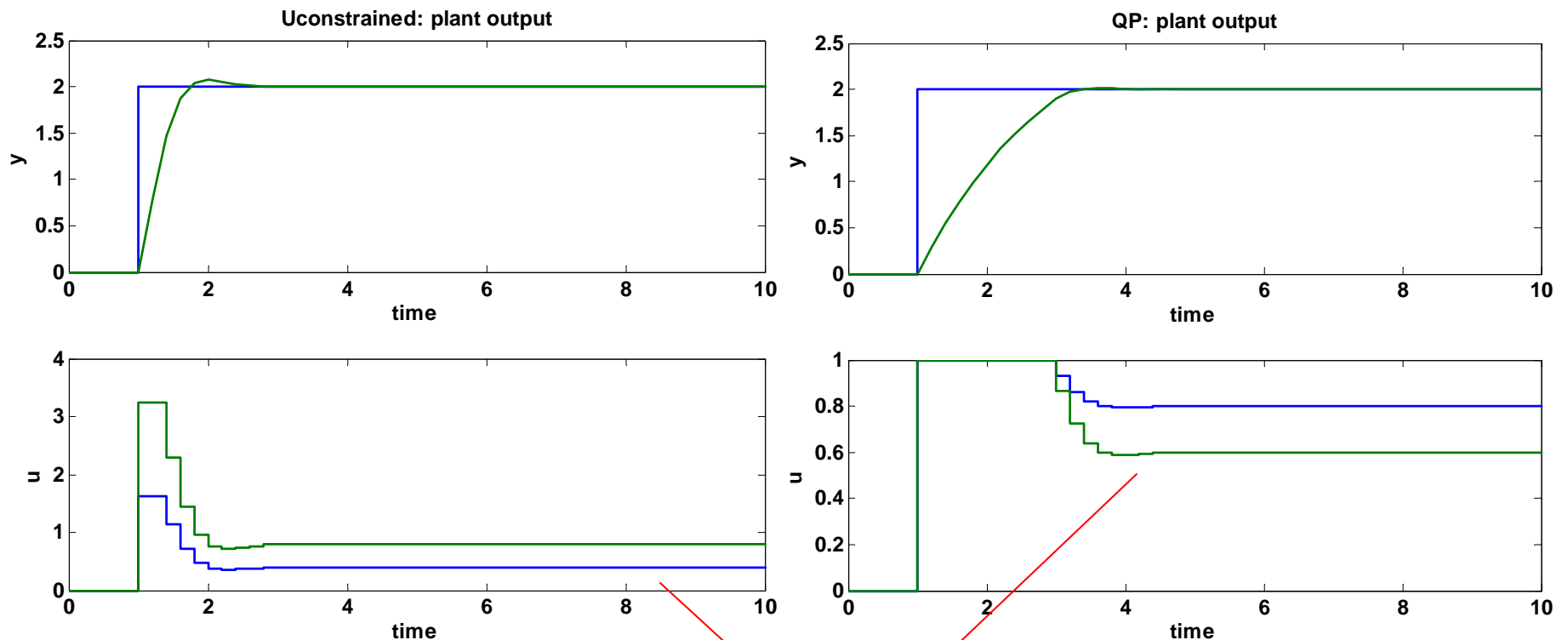
# Multivariable Examples

- Have shown clear advantages of KF-based MPC over DMC for SISO
  - Disturbance rejection
  - Unstable processes
- In general, MPC is especially powerful for handling constrained, multivariable systems (including “nonsquare”)
- Two “simple” example systems follow
  - Two-input, one-output (parallel valves)
  - Quadruple tank (two-input, two-output)

# 2 input – 1 output

$$y(s) = \frac{1}{2s+1}u_1(s) + \frac{2}{2s+1}u_2(s) + \frac{1}{2s+1}d(s)$$

$$W^u = \text{diag}(0.1, 0.1), W^y = 1, P = 10, M = 5$$



**Unconstrained**

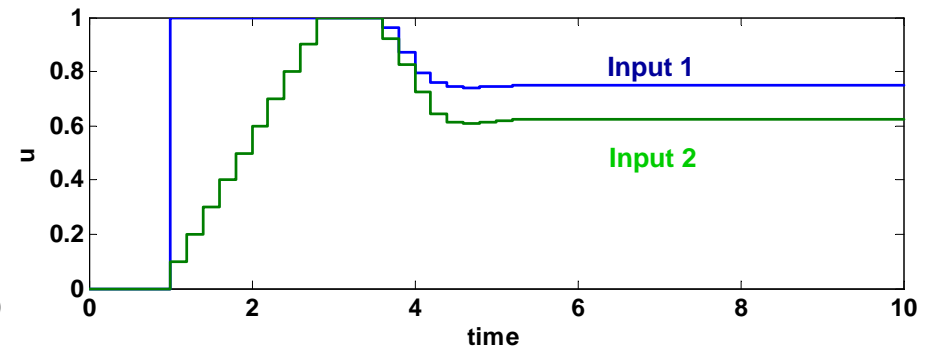
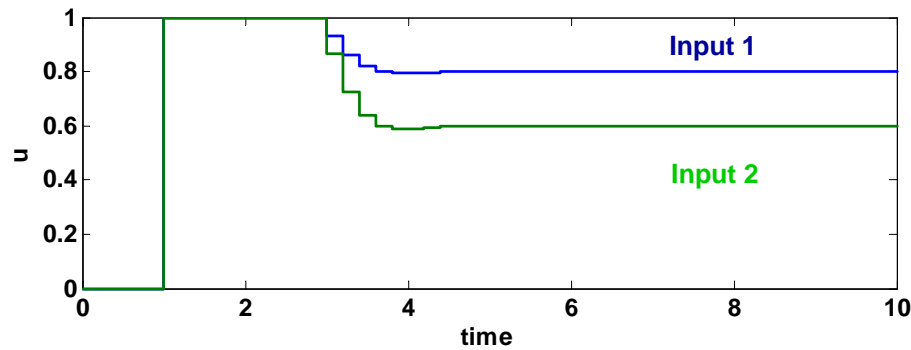
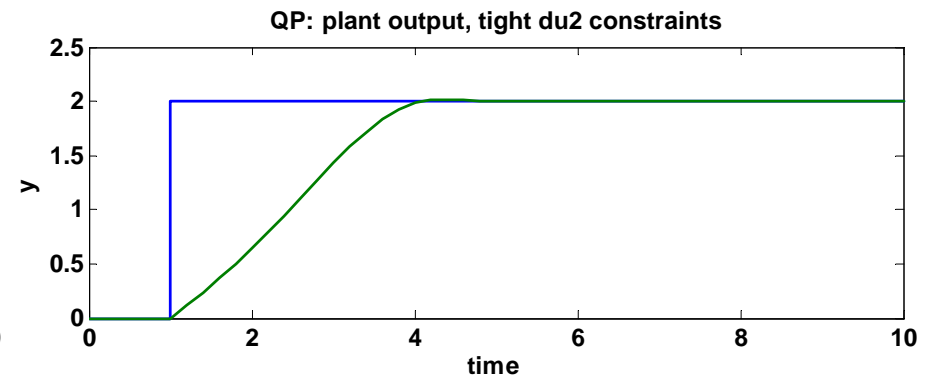
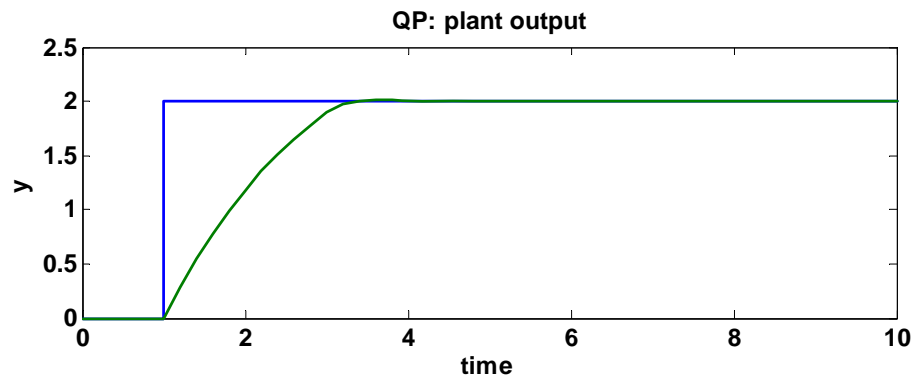
**Constrained**

**Inputs converge to different values**

B. Wayne Bequette

# 2 input – 1 output

## Setpoint Change, with constraints

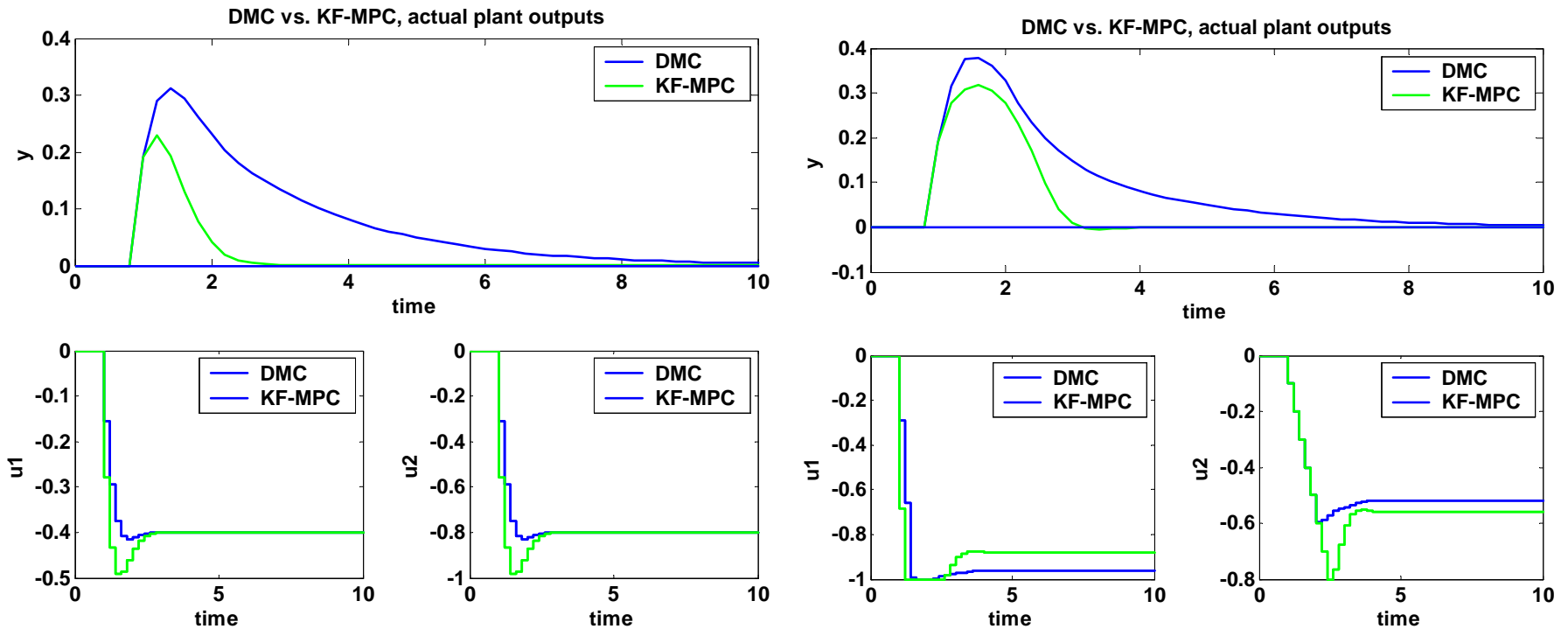


Constrained

Constrained,  $\Delta u_2 \leq 0.1$

# 2 input – 1 output

## Disturbance Rejection: unknown step in input

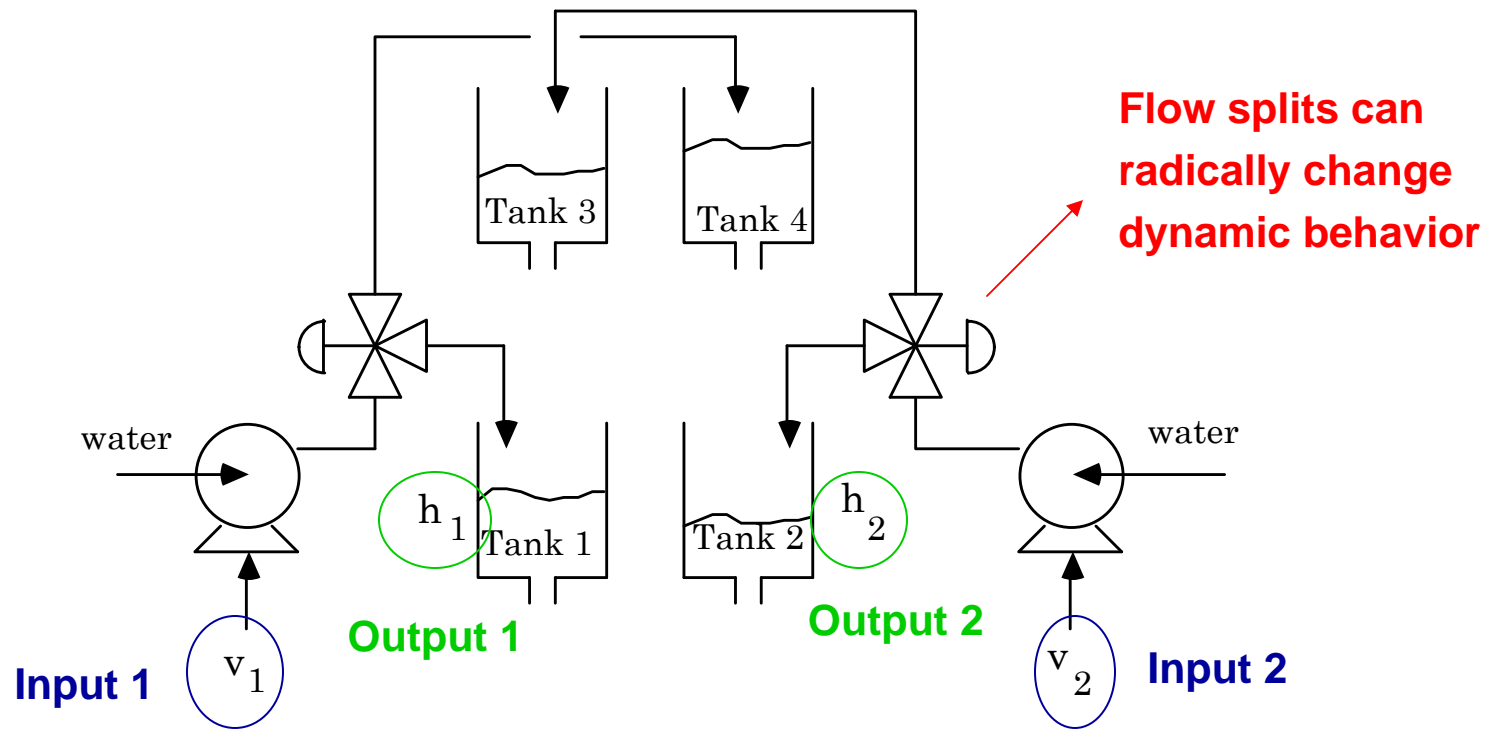


Tight velocity bounds (0.1) on  $u_1$

## 2 inputs – 1 output

- Many input combinations can achieve same output in steady-state
  - Tuning/constraints determine input values
- “Habituating Control” idea can be used
  - Fast, but “expensive” input for immediate disturbance rejection
  - Slow, but “cheap” input for long-term compensation

# Quadruple Tank Problem



# Quadruple Tank Problem

## Operating Point 1

$$G_1(s) = \begin{bmatrix} \frac{2.6}{62s + 1} & \frac{1.5}{(23s + 1)(62s + 1)} \\ \frac{1.4}{(30s + 1)(90s + 1)} & \frac{2.8}{(90s + 1)} \end{bmatrix}$$

## Operating Point 2

$$G_2(s) = \begin{bmatrix} \frac{1.5}{63s + 1} & \frac{2.5}{(39s + 1)(63s + 1)} \\ \frac{2.5}{(56s + 1)(91s + 1)} & \frac{1.6}{(91s + 1)} \end{bmatrix}$$

**No time-delays or right-half-plane zeros.**

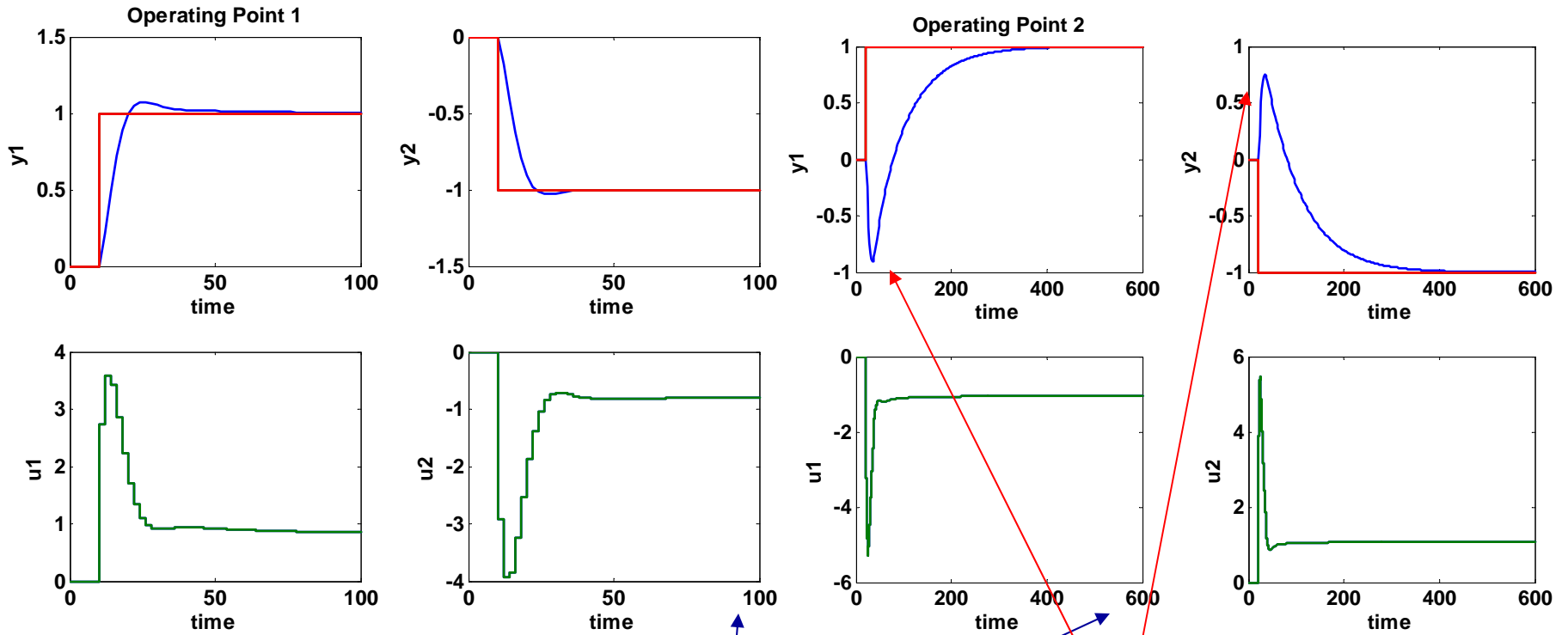
**Should we expect similar closed-loop performance?**



# Setpoint Responses

## Operating Point 1

## Operating Point 2



Note difference in time scales


“Wrong way” behavior

# Quadruple Tank Problem

## Operating Point 1 – Minimum Phase

$$G_1(s) = \begin{bmatrix} \frac{2.6}{62s+1} & \frac{1.5}{(23s+1)(62s+1)} \\ \frac{1.4}{(30s+1)(90s+1)} & \frac{2.8}{(90s+1)} \end{bmatrix} \quad z = -0.060 \text{ and } -0.018 \text{ sec}^{-1}$$

## Operating Point 2 – Nonminimum Phase (RHPT zero)

$$G_2(s) = \begin{bmatrix} \frac{1.5}{63s+1} & \frac{2.5}{(39s+1)(63s+1)} \\ \frac{2.5}{(56s+1)(91s+1)} & \frac{1.6}{(91s+1)} \end{bmatrix} \quad z = -0.057 \text{ and } +0.013 \text{ sec}^{-1}$$


# Multivariable Systems

- Can have right-half-plane “**transmission zeros**” even when no individual transfer function has a RHP zero
- Can have individual RHP zeros yet not have a RHPT zero
  - Fine performance when constraints are not active
  - May fail when one constraint becomes active or a loop is “opened”
- Can exhibit “directional sensitivity” – with some setpoint directions much easier to achieve than others
- Some of these MV properties cause challenges *independent of control strategy selected*

# State Space Form not Limiting

- Step Response or Discrete TF can be written in State Space form
- Virtually all recent theoretical results are based on a state space formulation

# Summary

- State Space MPC
- Disturbance Models
  - State/input or/and output disturbance
  - State/input form handles unstable systems
- Observer Design
  - Kalman Filter
- Can Control Outputs that are not measured
- Manipulated Input Blocking
  - Reduce number of decision variables

# Stability Proofs: Linear Models

- Unconstrained
  - Infinite horizon, State and Input Weights
    - LQ Control
    - Couple with Kalman Filter if some states are unmeasured
  - State penalty at end of horizon
- Constrained
  - Terminal state constraint
  - Stabilizing state feedback at end of prediction horizon